**Product Specification** 

Power Supply Supervisor + PWM

**SG6108** 

#### **FEATURES**

- PC half-bridge (or 494) power supply supervisor + PWM
- Highly Integrated and Few external components required
- Over voltage protection 3.3V/5V/12V (two 12V)
- Under voltage protection 3.3V/5V/12V (two 12V)
- Under voltage protection for -12V and/or -5V
- Over current protection 3.3V/5V/12V (two 12V)
- Over power protection
- Short circuit protection
- Power-down warning circuitry
- Power good circuitry
- Delay time for PSON and PG signals
- Remote ON/OFF function
- On-Chip oscillator and Error amplifier
- Latching PWM for cycle-by-cycle switching
- Soft-start
- Maximum 48% duty cycle

### **APPLICATIONS**

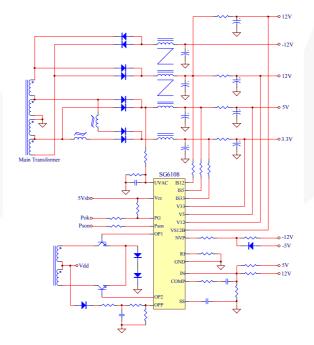
Switching mode power supplies for computers such as ATX, NLX, and SFX (micro-ATX).

#### **DESCRIPTION**

The SG6108 controller is designed for desktop PC switching mode power supplies. It provides everything necessary to monitor and control the output of the power supply. Remote ON/OFF control, power good circuitry, over-voltage protection, over-power protection, and over-current protection are all included. The SG6108 directly senses all the output rails for over voltage conditions without requiring external dividers. An innovative AC-signal sampling circuit provides an adequate power-down warning signal for PG. A built-in timer generates accurate timing for the control circuit.

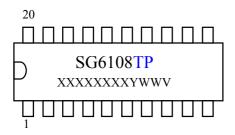
Cycle-by-cycle PWM switching protects the power transformer from saturation and ensures a rapid response for short-circuit protection. This greatly reduces the stress on the power transistors. The SG6108 utilizes a minimum number of external components while including all the functions needed for push-pull and/or half-bridge topologies. This decreases production costs and PCB space while increasing the power supply's MTBF.

### **TYPICAL APPLICATION**





## **MARKING DIAGRAMS**



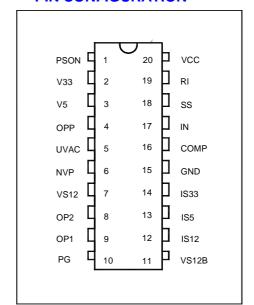
**T** : D = DIP

P: Z = Lead Free Null=regular package

XXXXXXXX: Wafer Lot

Y: Year; WW: Week V: Assembly Location

### **PIN CONFIGURATION**



## **ORDERING INFORMATION**

Part Number	Pb-Free	Package
SG6108DZ		20-Pin DIP

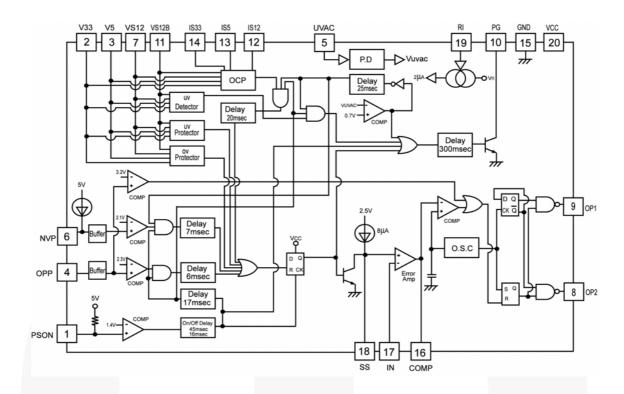
**SG6108** 

# **PIN DESCRIPTIONS**

Name	Pin	Type	Function
			Remote On/Off logic input for the CPU or the controller. Turns the PWM output on/off following a 45
PSON	1	I ania Immut	msec / 16 msec delay.
PSON	1	Logic Input	PSON = 0 means that the main SMPS is operational.
			PSON = 1 means that the main SMPS is off and the latch is reset.
V33	2	Analog Input	3.3V over-voltage/under-voltage control sense input.
V5	3	Analog Input	5V over-voltage/under-voltage control sense input.
OPP	4	Analog Input	Over-power sense input. This pin is connected to the driver transformer or the output of the current transformer. When not in use, this pin should be grounded.
UVAC	5	Analog Input	AC fail-detection. This detects main AC voltage under-voltage and/or failure.
NVP	6	Analog Input	The protection input for negative outputs, such as -12V and/or -5V. Trip voltage = 2.1V.
VS12	7	Analog Input	12V over-voltage/under-voltage control sense input.
OD2/OD1	0.40	1 1 0	The totem-pole output drivers of a push-pull PWM. The outputs are enabled (low) only when the
OP2/OP1	8/9	Analog Output	NAND gate inputs are high. The maximum duty cycle on an output (OP1 or OP2) is 48%.
PG	10	Logic Output	Power Good logic output, 0 or 1 (open-collector). PG = 1 means that the power is good for operation. The PG delay is 300 msec.
VS12B	11	Analog Input	Second 12V over-voltage/under-voltage control sense input.
IS12	12	Analog Input	12V over current protection sense input. The sink current = 8 x IRI
IS5	13	Analog Input	5V over current protection sense input. The sink current = 8 x IRI
IS33	14	Analog Input	3.3V over current protection sense input. The sink current = 8 x IRI
GND	15	Supply	Ground.
COMP	16	Analog Output	The error amplifier output and the input of the PWM comparator.
IN	17	Analog Input	The negative input of the error amplifier. The positive input of the error amplifier is a 2.5V reference voltage.
SS	18	Analog Input	Soft-start. This can be set through the external capacitor. The current source output at this pin is 8uA and the voltage is clamped at 2.5V.
			This is connected to the external resistor to provide a reference setting. It determines the PWM
D.	10	Program.	frequency, and also the OCP current setting. Setting RI = $75k\Omega$ will result in a PWM frequency of
RI	19	Analog Input	65kHz. Do not attach a capacitor to this pin. For OCP current settings please refer to the Over Current
			Protection section on page 13.
VCC	20	Supply	Supply voltage. $4.5 \text{V} \sim 5.5 \text{V}$ . It is connected to the 5V-standby.

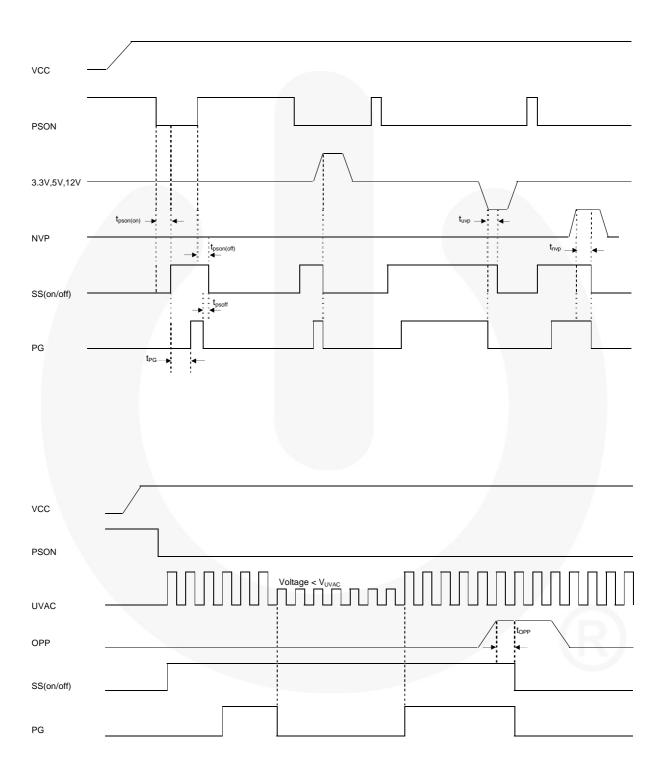


## **BLOCK DIAGRAM**



**SG6108** 

## **TIMING DIAGRAM**



**SG6108** 

# **ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Value	Unit
V <sub>CC, MAX</sub>	DC Supply Voltage at Pin 20	6	V
$I_{OUT}$	Input Current at PG,FB Pins	30	mA
V <sub>PIN, MAX</sub>	OP1, OP2, VS12, VS12B, IS12	-0.3 to 16	V
	OPP, UVAC, RI, SS, PG, IN, COMP, FB, NVP, IS5, V33, V5, PSON	-0.3 to 7	V
P <sub>D,25°C</sub>	Power Dissipation (T <sub>A</sub> =25°C)	1.5	W
P <sub>D,90°</sub> C	Power Dissipation (T <sub>A</sub> =90°C)	0.5	W
$T_{STG}$	Storage Temperature Range	-55 to +150	°C
$R_{\theta JA}$	Thermal Resistance, Junction-To-Case	28.58	°C/W
$T_{J,MAX}$	Operating Junction Temperature	150	°C
$T_{A,MAX}$	Operating Ambient Temperature	-20 to +85	°C
$T_{\rm L}$	Lead Temperature (Wave soldering or IR, 10seconds)	260	°C
	ESD Capability, HBM model	3.0	KV
	ESD Capability, Machine model	200	V

<sup>\*</sup> All voltage values, except the differential voltage, are given with respect to the network ground terminal.

## **RECOMMENDED OPERATING CONDITIONS**

Symbol	Parameter	Value	Unit
$V_{CC}$	DC Supply Voltage	4.5 to 5.5	V
$T_A$	Operating Ambient Temperature	-20 to +85	°C

<sup>\*</sup> For proper operation

# Electrical Characteristics (VCC=5V, T<sub>A</sub> = 25°C,RI=75Kohm,unless noted)

# **VCC Section**

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
$I_{CC1}$	Total Supply Current	PSON=Low,			10	mA
		OP1/OP2 1000pF				
$I_{CC2}$	Total Supply Current	PSON=High,	- 1	5	10	mA
		OP1/OP2 1000pF				

<sup>\*</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device.



## **Protection Section**

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
$V_{OVP1}$	Over-Voltage Protection 3.3V		3.9	4.1	4.3	V
$V_{OVP2}$	Over-Voltage Protection 5V		5.8	6.1	6.5	V
V <sub>OVP3</sub>	Over-Voltage Protection VS12,VS12B		13.75	14.05	14.35	V
$V_{ m UVP1}$	Under-Voltage Protection 3.3V		2.0	2.6	2.8	V
$I_{ m UVP2}$	Under-Voltage Protection 5V		3.0	3.6	3.9	V
V <sub>UVP3</sub>	Under-Voltage Protection VS12,VS12B		6.0	7.2	8.0	V
$J_{\rm UVS1}$	Under-Voltage Sense 3.3V for PG Low		2.5	2.8	3.0	V
$I_{\rm UVS2}$	Under-Voltage Sense 5V for PG Low		4.0	4.3	4.5	V
UVS3	Under-Voltage Sense VS12,VS12B for PG Low		9.4	10.1	10.4	V
7 <sub>OPP</sub> * 1	Over-Power Protection (With T <sub>OPP</sub> delay time)	$V_{UVAC} = 1.5V$	2.23	2.3	2.37	V
V <sub>OPPH</sub>	Over-Power Protection (Without delay time)		3.0	3.2	3.4	V
$I_{\rm X}$	Disable Under Voltage / Over Power Protection threshold		0.2	0.3	0.4	V
NVP	Negative Voltage Protection: Voltage Level		2.0	2.1	2.2	V
IVP	Negative Voltage Protection: Source Current		57	64	72	uA
UVEDA	Under Voltage Enable Delay Time	VS12/B = 0.8V	10	17	22	mS
UVEDB	Under Voltage Enable Delay Time	VS12/B = 5V	22	30	40	mS
OVP	Timing for Over-Voltage Protection		0.37	0.7	1.35	mS
UVP	Timing for Under-Voltage Protection		1.2	2.7	4.2	mS
UVS	Timing for Under-Voltage Sense (PG Low)		0.9	1.6	2.2	mS
OPP	Timing for Over-Power-Protection		4	6	8	mS
NVP	Timing for Negative Voltage Protection		3.3	7	10.2	mS
OCP	Timing for Over-Current Protection		15	20	25	mS
,	Timing for Over County Destrotion Founding Funds	SS go Hi, VS12/B=0.8V	10	17	22	mS
OCP(EN)	Timing for Over-Current Protection Function Enable	SS go Hi, VS12/B =5V	22	30	40	ms
OFFSET	OCP Comparator Input Offset Voltage		-5		5	mV
REF	Ratio of IS33, IS5, IS12 to RI Current Setting		7.6	8	8.4	
.K33	V33 Leakage Current	V33 = 3.3V			50	uA
.K5	V5 Leakage Current	V5= 5V			50	uA
LK12AB	VS12, VS12B Leakage Current	VS12 =12V VS12B = 12V		/	50	uA

Note  $^{*1}$ :  $V_{OPP} = (2/3) \times V_{OPP} + (1/3) \times V_{UVAC}$ ;

# **Power Good Section**

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
$T_{PG}$	Timing for PG Delay		200	300	400	mS
$T_{UVAC}$	Timing for PG Fail	UVAC to PG	75	150	300	uS
$V_{UVAC}$	UVAC Voltage Sense for PG		0.68	0.7	0.72	V
V <sub>UVAC(DIS)</sub>	Disable UVP,NVP,OCP Voltage		0.68	0.7	0.72	V
T <sub>UVAC(DIS)</sub>	Timing for Disable UVP,NVP,OCP	$UVAC < V_{UVAC(DIS)}$	75	150	300	uS
T <sub>UVAC(EN)</sub>	Timing for Enable UVP,NVP,OCP	$UVAC > V_{UVAC(DIS)}$	15	25	35	mS
T <sub>R</sub> *2	PG Output Rising Time	CL =100pF Pull 2.2K to 5V		1	3	uS
T <sub>F</sub> *2	PG Falling Time	CL =100pF Pull 2.2K to 5V		300	500	nS
$V_{OL2}$	Power Good Output Saturation Level	$I_{PG} = 5 \text{mA}$			0.5	V
$I_{IO2}$	Power Good Leakage Current	$V_{PG} = 5V$			1	uA

Note \*2: Not tested in production test;

**SG6108** 

## **Remote On/Off**

Symbol	Parameter	Test Cor	ndition	Min.	Тур.	Max.	Unit
V <sub>PSON</sub>	PSON Input Threshold Level			1	1.4	2.0	V
$I_{PSON}$	Remote Input Driving Current					0.5	mA
	PSON to On/Off Time						
T <sub>PSON(ON)</sub>	On			25	45	65	mS
T <sub>PSON(OFF)</sub>	Off			8	16	24	
T <sub>PSOFF</sub>	PG Low to Power Off Time			1.5	2	6.3	mS

# **Error Amplifier**

Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
$V_{2.5}$	Reference Voltage		2.45	2.5	2.55	V
${\rm I_{IB}}^{*2}$	Input Bias Current				0.1	uA
V <sub>DEV,T2.5</sub> *2	Deviation of V <sub>2.5</sub> Over Temperature			10	30	mV
$I_{COSI}$	I <sub>COMP</sub> Source Current	IN Short COMP	0.4			mA
		VCOMP=2.4V				
I <sub>COSO</sub>	I <sub>COMP</sub> Sink Current	IN Short COMP	1.2			mA
		VCOMP=2.6V				
A <sub>VOL</sub> * 2	Open-Loop Voltage Gain		50	60		dB
BW *2	Unity Gain Bandwidth		0.3	1		MHz
PSRR *2	Power Supply Rejection Ratio		50			dB

# **Oscillator Section**

Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
Fosc	PWM Frequency		60	65	70	KHz
$DC_{MAX}$	Max Duty Cycle		41	44.5	48	%
$R_{\rm I}$	RI Resistor Range		70		105	ΚΩ
V <sub>REFRI</sub>	RI Reference Voltage	V	0.98	1	1.02	V
$T_{DEAD}$	PWM Dead Time		1.3	1.7	2.1	uS

# **Soft-Start Section**

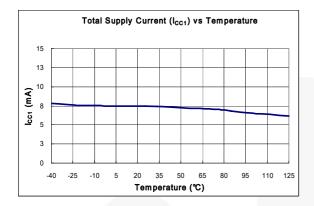
Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
$I_{SS}$	Charge Current		6.7	8	9.3	uA
I <sub>SSINK</sub>	Sink Current		1			mA

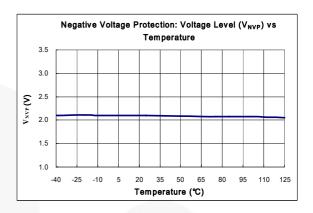
# **PWM Output Section (OP1, OP2)**

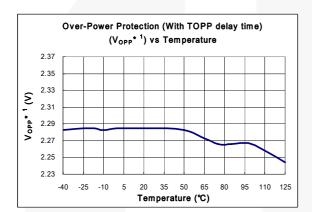
Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
$V_{OL}$	Output Voltage Low	Io = 20mA			1	V
$V_{\mathrm{OH}}$	Output Voltage High	VDD = 5V	4			V
$R_{O}$	Output Impedance of V <sub>OH</sub>		0.8		2.6	kΩ

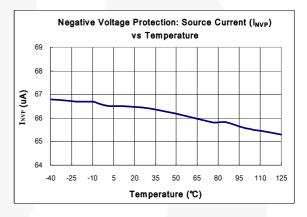


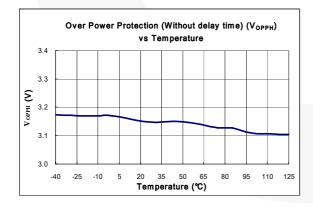
### **TYPICAL CHARACTERISTICS**

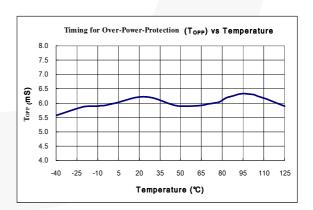




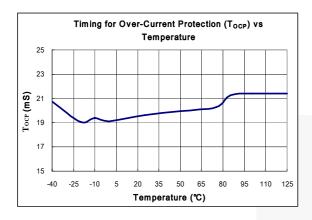


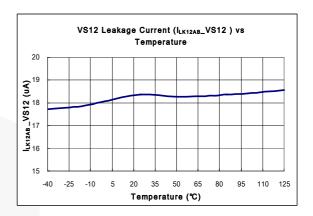


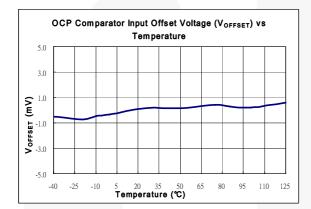


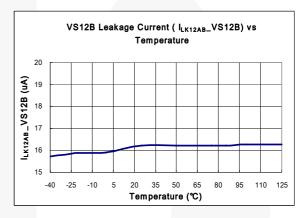


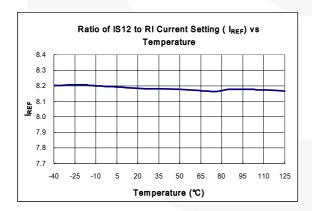


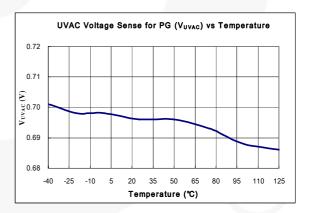




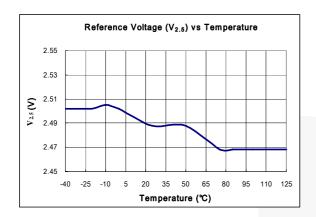


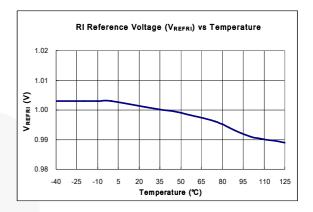


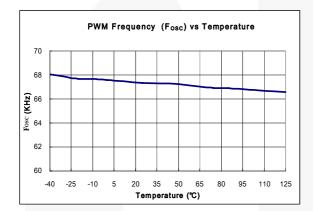


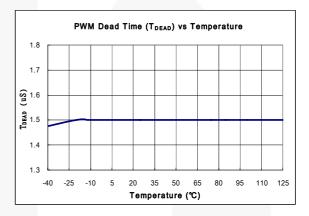


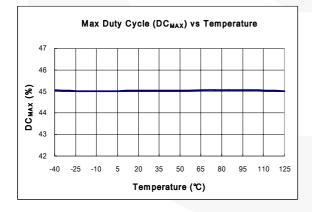


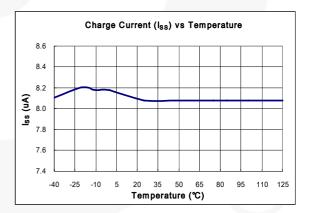














### **Operation description**

Protection against over-voltage, short-circuit and fault conditions is mandatory for PC power supplies. Many commonly used protection circuits require large numbers of discrete components and comparators which occupy significant PCB space and add to assembly costs. This single chip controller IC provides complete protection, shunt regulation, and PWM control with comparatively few components. The SG6108 is an ideal controller IC for PC switching mode power supplies. The features and benefits of this device are listed on the first page.

#### **Feature Notes**

- Over voltage protection can be implemented without any additional components. Over-voltage sense levels are 3.3V/5V/12V or 4.1V/6.1V/14.05V.
- The SG6108's over current protection monitors the 3.3V, 5V, and the two 12V levels. The voltage difference across the external current shunt is used to provide accurate OCP.
- Over current and/or short circuit protection can also be implemented by using over-power protection. To do this, the OPP pin must be connected to the current transformer (driver transformer).
- The power good signal is asserted to indicate that the 3.3V/5V/12V signals are above the under voltage threshold level. The PG pin goes high when the above condition is reached. A 2kΩ pull up resistor may be optionally connected to the 5V lead.
- VCC can be supplied from the 5V-standby. Whenever the VCC voltage exceeds 7V the entire SG6108 will be shutdown and reset (except for the shunt regulator). No extra power supply is needed.
- Complete PWM control circuitry includes an error amplifier for push-pull or half-bridge operation.

### **Supervisory Circuit Operation**

The PC generates the remote ON/OFF logic (PSON) signal. When this is "Low" the power supply is "On". When it is "High", the power supply is switched off. The remote ON/OFF signal is supplied to the PSON input.

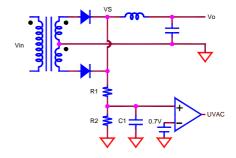
### **Application Guides**

The application guide shows the key features of the SG6108 and illustrates how to use it in an ATX switching mode power supply (SMPS). The SG6108 is suitable for half-bridge, push-pull topologies. The PWM section of the SG6108 comprises a built-in 65kHz oscillator and high immunity circuits, which protect the system from noise interference and provide higher noise margins against improper PCB layout. The SG6108 has OVP, UVP, and OCP for the 3.3V and 5V signals, as wells as both 12V signals. NVP is used for negative voltage protection (-12V/-5V). UVAC (AC failure detection) is used to measure the AC line voltage.

#### **AC Failure Detection**

UVAC is connected to the secondary power transformer through a resistor divider. UVAC is used for detecting the AC line voltage. Whenever the UVAC voltage drops below 0.7V for a period of time such as 150usec, the PG (power good) signal will be pulled low to indicate an AC line power-down condition.

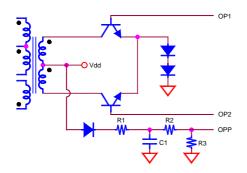
The voltage amplitude of the PWM switching signal in the secondary power transformer is proportional to the AC line voltage. The power-down warning threshold can be controlled by adjusting the resistor-divider ratio. A small



capacitor must be connected between UVAC and ground to filter the switching noise.

### **Over Power Protection (OPP)**

OPP (over power protection) is used to detect over power and short-circuit conditions. Whenever the  $V_{OPP}$  voltage (p.7, Note-1) exceeds 2.3V for longer than 6msec the SG6108 will pull PG low and lock off the power outputs.



# **Over Current Protection (OCP)**

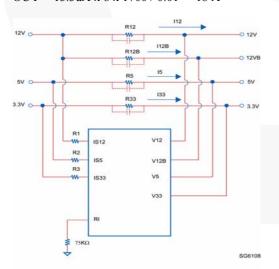
The SG6108 provides over current protection for the 3.3V, 5V and the two 12V rails. Whenever an OCP condition occurs at any of the voltage rails, PG will be low and PWM will be turned off. The internal OCP comparators have a very small offset voltage (±5mV).

The sink currents of IS33, IS5, and IS12 are 8 times the current at the RI pin. The current at the RI pin is  $V_{RI}/RI$ . Here is an example demonstrating how to set the over current protection threshold:

$$I_{RI} = V_{RI}/RI = 1.0V/75k = 13.3uA$$

If R1 =  $1700\Omega$  and R12 =  $0.01\Omega$ , then

OC 
$$P = 13.3uA \times 8 \times 1700 / 0.01 = 18 A$$

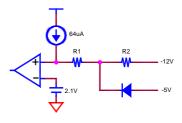


## **Negative Voltage Protection**

NVP provides UVP (under voltage protection) for negative voltage outputs. An under-voltage condition indicates an overload at a negative voltage output. For example, the -12V output may drop to -10V during overload situations. The resistance of a resistor connected from NVP to the negative voltage output determines the protection threshold voltage. NVP outputs a 64uA constant current to the negative voltage output via this resistor. Whenever the NVP voltage exceeds 2.1V for longer than 7 msec the SG6108 will lock the power outputs off.

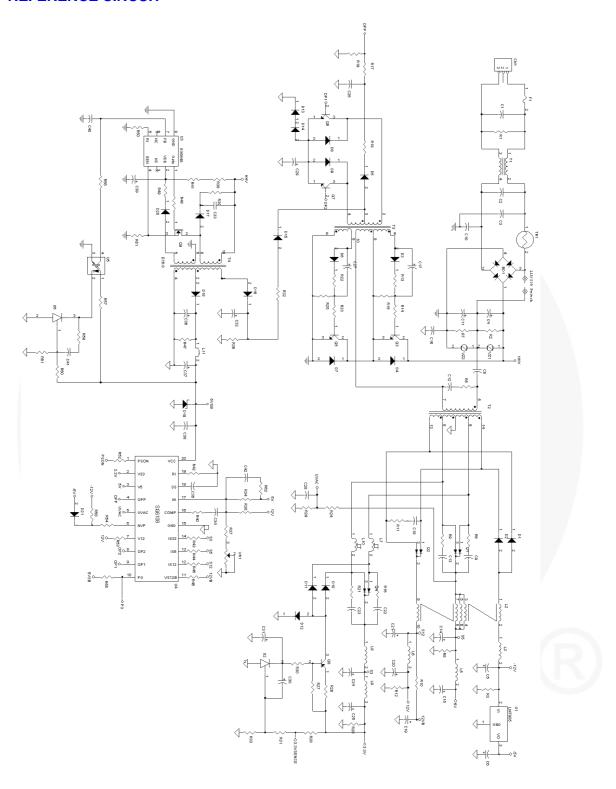
$$VNVP = 64 \text{ uA} \times (R1 + R2) - V-12V$$

The power outputs will be locked off when  $VNVP \ge 2.1V$ .



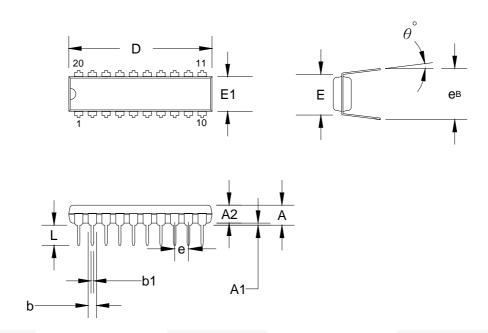


## **REFERENCE CIRCUIT**





# PACKAGE INFORMATION 20 PINS - DIP (D)



# **Dimensions**

Symbol	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			5.334			0.210
A1	0.381			0.015		
A2	3.175	3.302	3.429	0.125	0.130	0.135
b		1.524			0.060	
b1		0.457			0.018	
D	24.892	26.162	26.924	0.980	1.030	1.060
E		7.620			0.300	
E1	6.223	6.350	6.477	0.245	0.250	0.255
e		2.540			0.100	
L	2.921	3.302	3.810	0.115	0.130	0.150
$e_{\mathrm{B}}$	8.509	9.017	9.525	0.335	0.355	0.375
θ•	0°	7°	15°	0°	7°	15°







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